

Freq LT $\rightarrow H(s)$

p/z: if some p has ≥ 0 real part, unstable

z: no stability eff, changes I/O effect

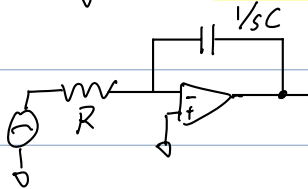
$H(s) \rightarrow$ impulse/step resp, natural/forced, zi/zs

$\rightarrow H(j\omega) \rightarrow$ bode plt

$$x(t) = A \sin(\omega t)$$

$$y(t) = |H(j\omega)| \sin(\omega t + \angle H(j\omega))$$

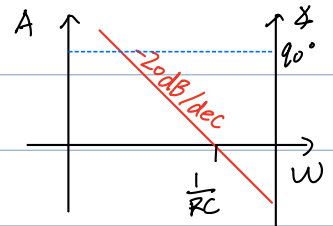
E. J



$$V_{out}(s) = 0 - \frac{1}{sC} \cdot V_{in} \cdot \frac{1}{R}$$

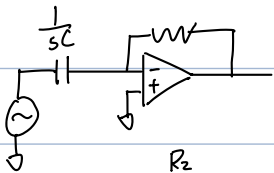
$$H(s) = -\frac{1}{sRC}$$

$$H(j\omega) = \frac{j}{\omega RC}$$



p: $s=0$

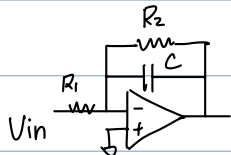
E. $\frac{d}{dx}$



$$H(s) = -sRC$$

+20 dB/dec, -90°

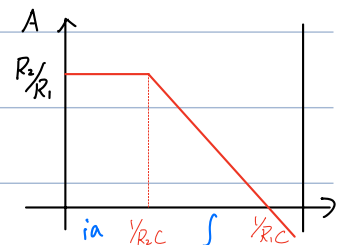
E. LPF



$$H(s) = -\frac{R_2}{R_1} \frac{1}{1+sR_2C} \quad p: s = -\frac{1}{R_2C}$$

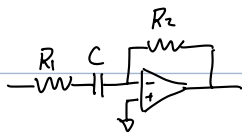
ia $\rightarrow$ ideal $\int$

$-180^\circ \rightarrow -270^\circ (+90^\circ)$



z: $s=0$

E. HPF



$$H(s) = -\frac{R_2}{R_1} \frac{sR_1C}{1+sR_1C}$$

ideal $\frac{d}{dx} \rightarrow$ ia

$-90^\circ \rightarrow -180^\circ$

1p/z

E. BPF (V)

$$H(s) = -\frac{R_2}{R_1} \frac{sR_1C_1}{1+sR_1C_1} \frac{1}{1+sR_2C_2}$$

$-90^\circ \rightarrow -180^\circ \rightarrow -270^\circ$

E. HW2 All comp. give same pole, but diff. zero pf.

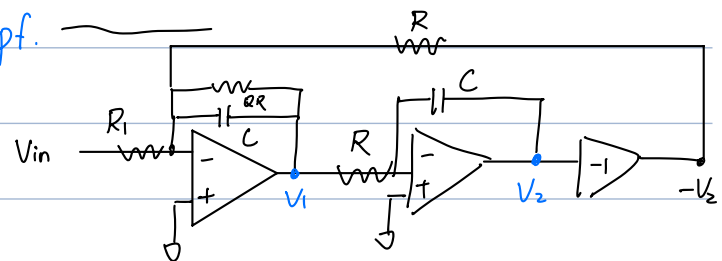
$$\frac{V_{in}}{R} + \frac{V_1}{QR} + sC V_1 - \frac{V_2}{R} = 0$$

$$sC V_2 + \frac{V_1}{R} = 0$$

$$\begin{bmatrix} \frac{1}{QR} + sC & -\frac{1}{R} \\ \frac{1}{R} & sC \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} -\frac{1}{R_1} \\ 0 \end{bmatrix} V_{in}(s)$$

$$V_1(s) = \frac{\begin{vmatrix} -\frac{1}{R_1} V_{in} & -\frac{1}{R} \\ 0 & sC \end{vmatrix}}{\begin{vmatrix} \frac{1}{QR} + sC & -\frac{1}{R} \\ \frac{1}{R} & sC \end{vmatrix}} = \frac{-\frac{1}{R_1} sC V_{in}}{sC(\frac{1}{QR} + sC) + \frac{1}{R^2}} = \frac{-\frac{R_1}{R} sRC V_{in}}{1 + s\frac{RC}{Q} + s^2 R^2 C^2}$$

2p/z BPF



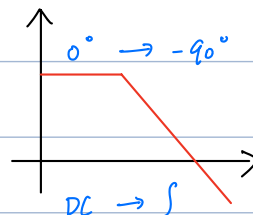
$$V_2(s) =$$

2p LPF

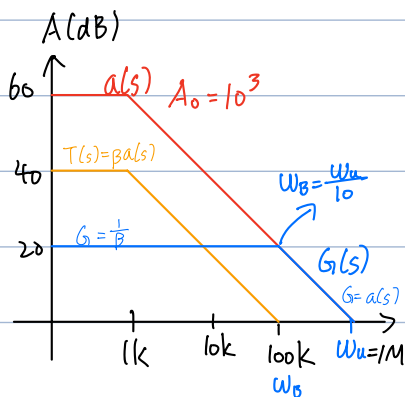
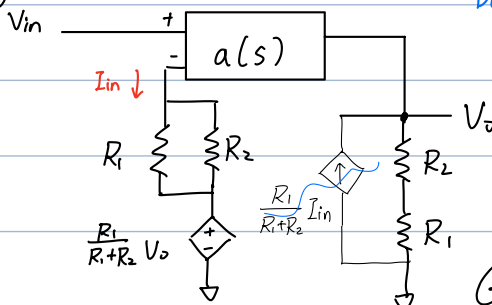
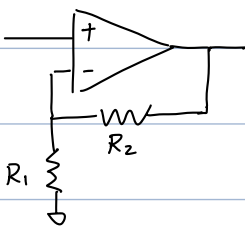
Opamp (bw)

Single pole $A(s) = \frac{A_0}{1 + \frac{s}{\omega_u/A_0}} \rightarrow$ corner at $\frac{\omega_u}{A_0}$; unity at ω_u

ω_u : GBW (if 1st order up to ω_u)



E.ni



$$\beta = \frac{1}{10}, T(s) = \beta a(s)$$

$$G = \frac{1}{\beta} \frac{T(s)}{1+T(s)}$$

$$T(s) \gg 1, G = \frac{1}{\beta} \rightarrow \text{BW} \uparrow$$

$$T(s) < 1, G = a(s) \text{ (roll-off)}$$

$$G(s) = \frac{a(s)}{1 + \beta a(s)}$$

$$= \frac{A_0}{1 + s \frac{A_0}{\omega_u} + \beta A_0}$$

$$= \frac{A_0}{(1 + \beta A_0) + s \frac{A_0}{\omega_u}}$$

$$= \frac{1}{1 + \beta A_0} \frac{1}{1 + \frac{s}{\omega_B}} \quad (\omega_B = \omega_u \frac{1 + \beta A_0}{A_0})$$

$$\approx \frac{1}{\beta} \frac{1}{1 + s/\omega_B}, \omega_B = \frac{\omega_u}{1/\beta}$$

$$Z_{out}(s) = \frac{R_{out}}{1 + T(s)} = \frac{r_o \parallel (R_1 + R_2)}{1 + T(s)}$$

Model w/ L ($Z \uparrow$ as $\omega \uparrow$) in sr.

$$E. R_{out} = 100 \Omega; Z_{out}: (1 \Omega \rightarrow 100 \Omega)$$

$$L = \frac{r_o}{\omega_B} = \frac{r_o}{\beta \omega_u} \approx 160 \mu H$$

If Z_L is cap $\rightarrow$ resonate: (

10k · 100

$$Z_{in}(s) = (R_1 \parallel R_2) + r_{in}(1 + a(s)\beta)$$

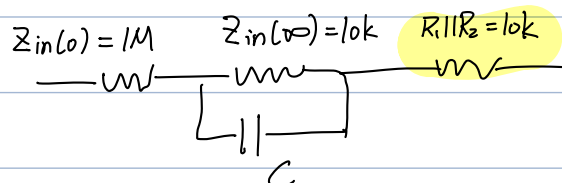
only r_{in} !

$$E. R_{in} = 10k, R_1 \parallel R_2 = 10k$$

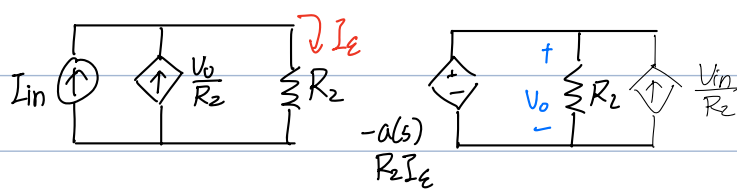
$$Z_{in}(s) = 10k + 10k(1 + a(s)\beta)$$

$$Z_{in}(0) = 1 M\Omega, Z_{in}(\infty) = 20k$$

$$Z_c = \frac{1}{j\omega C} \quad C \approx 0.16 nF$$



E. IA (sh-sh)



$$A(s) = \frac{V_o}{I_e} = -R_2 a(s)$$

$$\beta = -\frac{1}{R_2}$$

$$T = \beta A = a(s)$$

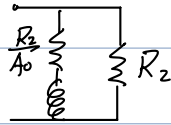
$$G = \frac{A}{1+\beta A}$$

If R_s is present in src $\rightarrow$ G - BW tradeoff

$$Z_{in} (\text{assume } r_{in} = \infty) = \frac{R_{in}}{1+T(s)} \quad (\text{Miller effect})$$

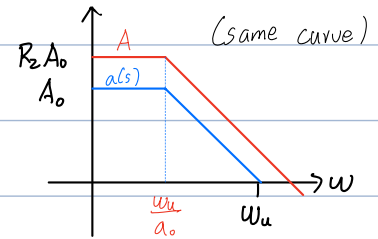
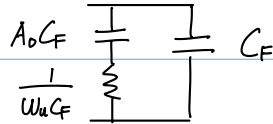
$$R_{in} = R_2$$

$$L_{eq} = \frac{R_2}{\omega_u}$$



$$Z = j\omega L$$

$$\text{If } Z_f = \frac{1}{sC_F}$$



$$a(\omega_u) = 1$$

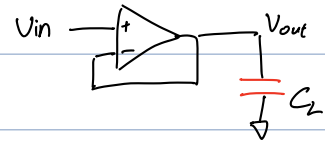
$Z_{out}??$

Stability

E unity buf. $\beta=1$, $\omega_B = 1\text{MHz}$

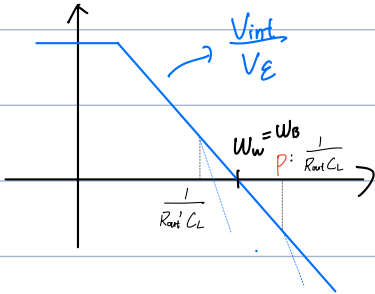
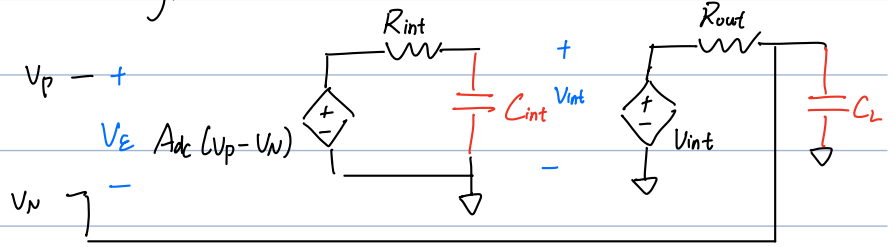
If $C_{out} \uparrow \rightarrow$ ringing! More severe if $C_L \uparrow$

If $\omega_B \uparrow \rightarrow$ ringing $\uparrow$ (faster, but more ring)

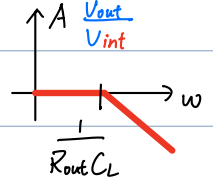


Two poles: C_{int} and C_L

two stages to isolate



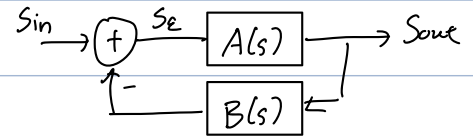
additional pole at $\omega = \frac{1}{R_{out} C_L}$



$$G(s) = \frac{A(s)}{1 + \beta(s)A(s)} \approx \begin{cases} \frac{1}{\beta(s)} & \text{for } A\beta \gg 1 \\ A(s) & \text{for } A\beta \ll 1 \end{cases}$$

(fine if $A\beta \neq -1$)

Let $A(s) = \frac{N(s)}{D(s)}$ $\rightarrow G(s) = \frac{N(s)}{D(s) + \beta N(s)}$ (assume $\beta(s) = \beta$)

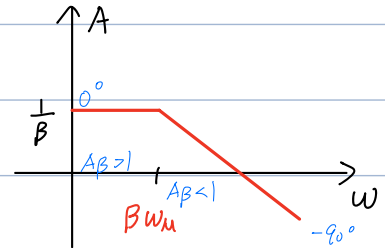


zero of $A(s) \rightarrow z$ of $G(s)$

pole of $A(s) \rightarrow ??$ matters for stability

1st order

E. $\beta(s) = \beta$, $A(s) = \frac{\omega_u}{s} \rightarrow G(s) = \frac{\omega_u}{s + \beta\omega_u}$



$p = -\beta\omega_u < 0 \rightarrow$ stable

Second order E. $A(s) = \frac{A_0}{(s/\omega_{p1} + 1)(s/\omega_{p2} + 1)}$

$$G(s) = \frac{A(s)}{1 + \beta A(s)} = \frac{A_0 \omega_{p1} \omega_{p2}}{s^2 + (\omega_{p1} + \omega_{p2})s + (1 + A_0\beta)\omega_{p1}\omega_{p2}}$$

$$\equiv \frac{A_0}{s^2 + \left(\frac{\omega_u}{Q}\right)s + \omega_u^2}$$

Let $\omega_u^2 = (1 + A_0\beta)\omega_{p1}\omega_{p2}$

$Q = \frac{\sqrt{(1 + A_0\beta)\omega_{p1}\omega_{p2}}}{\omega_{p1} + \omega_{p2}}$

If $\omega_{p2} \gg \omega_{p1}$, $A\beta \gg 1$, $Q \approx \sqrt{\frac{A_0\beta\omega_{p1}}{\omega_{p2}}} \equiv \sqrt{\frac{\omega_u}{\omega_{p2}}}$

We typically want $Q < 1$, so $\omega_{p2} > \omega_u$ (worst case unity)

$Q = 0.5$: critically damped $\omega_u = 0.25\omega_{p2}$

under

Freq domain criteria $\frac{A(s)}{\beta(s)A(s)}$ stability given $\beta(s)A(s)$

$G(s) = \frac{A(s)}{1 + \beta(s)A(s)} \rightarrow p$ at $A(s)\beta(s) = -1$, stable if all poles in LHP

1. ω_c where $\angle[A(j\omega_c)B(j\omega_c)] = -180^\circ \Rightarrow |A(j\omega_c)B(j\omega_c)| < 1$

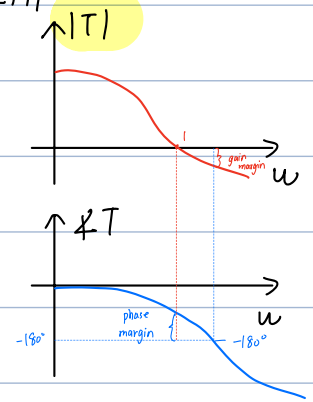
At phase crossover, $|T| < 1$ (gain margin $\frac{1}{|T|} > 1$)

2. ω_u where $|A(j\omega_u)B(j\omega_u)| = 1 \Rightarrow \angle(\text{---}) > -180^\circ$

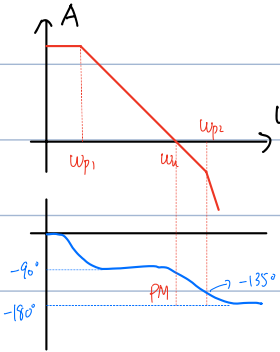
At gain crossover, $\angle T > -180^\circ$ (phase margin $180^\circ + \angle T > 0$)

if too close to margin, risk unstable

ω_c and ω_u need to be uniquely defined. Else need Nyquist test



E.



If add p_3 , it will drag down $\angle \rightarrow PM \downarrow$

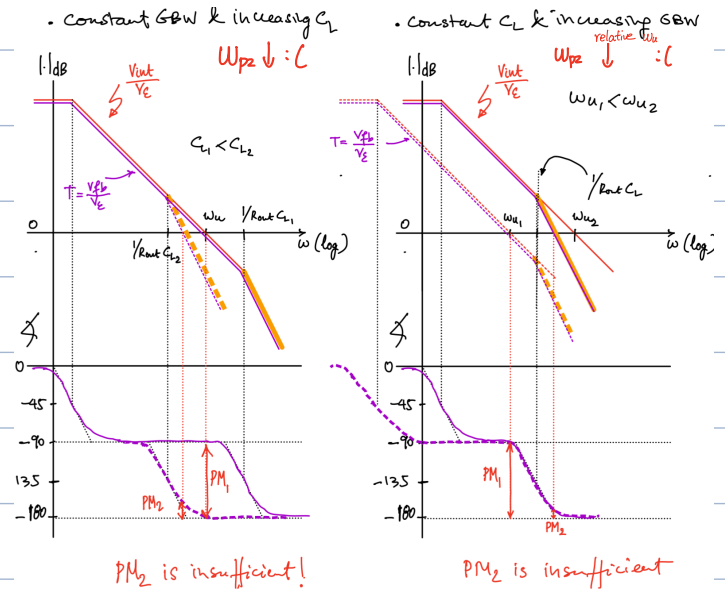
If $\beta \downarrow$, $T(s) \downarrow$, crossover at lower $\omega \rightarrow PM \uparrow$

$C_L \uparrow$ pole $\frac{1}{R_{out}C} \downarrow$ (both A and $\angle$) $\rightarrow PM \downarrow$

Aim for $PM > 45^\circ$, $60^\circ \rightarrow$ no ringing, critically damped

Need $\omega_u = A_0 \omega_{p1} < \omega_{p2} < \omega_{p3}$ (higher p no interfere)

if $\beta < 1$, $T(s) = \beta A(s) \downarrow \rightarrow \omega_{p3} \downarrow \rightarrow pm \uparrow$

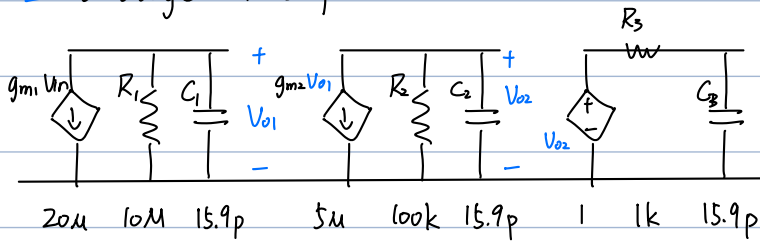


Compensation

change $\beta(s)$ if opamp given

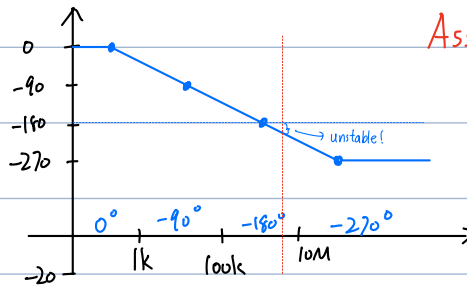
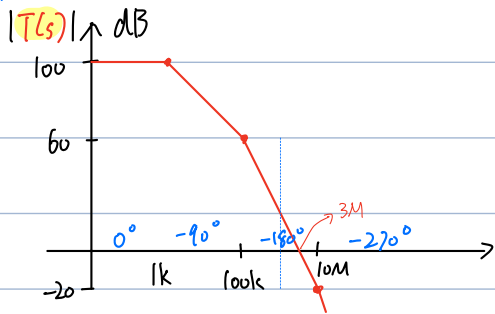
change $A(s)$ if designing opamp

E. 3-stage w/ 3 poles



DC $A_1 = -200$ $A_2 = -500$ $A_3 = 1$ $A = 100k$ (100dB)

p (Hz) -1kHz -100k -10M BW = 1kHz



Assume $\beta = 1$ (worst case)

Unstable: C

Sln. (can't change cap) Need $\omega_u = \omega_{p2} \rightarrow 45^\circ$ pm

1. E Reduce gain @ p_2 to 0dB $\rightarrow$ reduce A by 60dB

$\therefore$ Less benefit from feedback E. accuracy, Z_{in}/Z_{out}

2. E. add Cap. (dominant pole compensation) Make $C_1 = 15.9$ nF $\rightarrow p_1 = 1$ Hz (pull p_1 low)

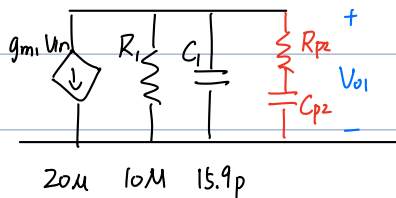
$$\omega_u = \omega_{p2} = 100k \text{ Hz} \rightarrow \omega_{p1} = \frac{100k}{10^{100/20 \text{ dB}}} = 1 \text{ Hz}$$

pm $\uparrow$, bw $\downarrow$:C

* Moving p_2 won't help. Need all other poles beyond $\omega_u = A_0 \omega_d$

3. p/z compensation Add z to cancel p₂.

(R dom. below p ; C above).



$$\omega_{p1} = -\frac{1}{R_1(C_1 + C_{p2})} \quad R_{p2} \approx 0 \quad \text{Assume } R_{p2} \ll R_1$$

$$\omega_z = -\frac{1}{R_{p2} C_{p2}} \quad C_{p2} \gg C_1$$

$$\omega_{p4} = -\frac{1}{R_{p2} C_1} \quad R_1 \approx \infty \quad \omega_{p1} \ll \omega_{p2}$$

Also add zeroes to cancel poles

Problem: if p/z misalign $\rightarrow$: (more overshoot, -)

E. $R_{p2} = 500k$, $C_{p2} = 500p$

$$\omega_4 = GBW = \omega_1 a_0 \rightarrow \frac{1}{R_{p2} C_1} = a_0 \frac{1}{R_1 (C_{p2} + C_1)}$$

$$\omega_z = \frac{1}{R_{p2} C_{p2}} = \omega_z$$

$$C_{p2} = \sqrt{\frac{a_0 C_1}{R_1 \omega_z}} = 500p$$

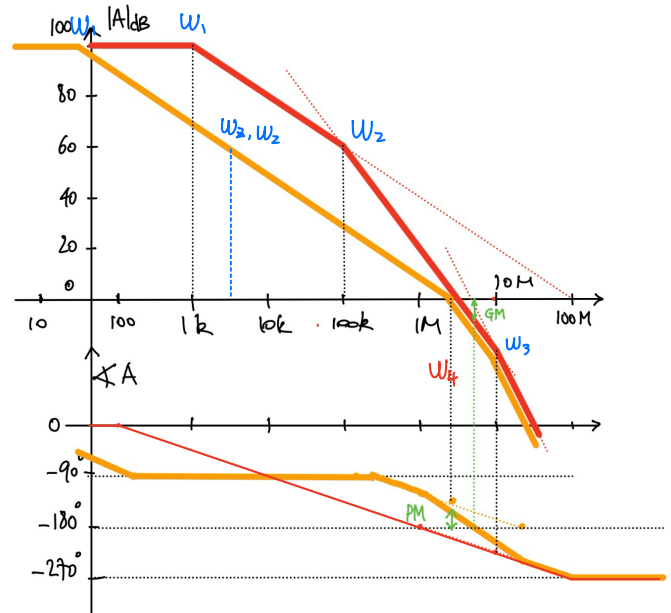
$$R_{p2} = \frac{1}{\omega_z C_{p2}} = 3.2k$$

$$\omega_1 \approx 31Hz$$

$$\omega_4 \approx 31MHz \rightarrow 45^\circ pm$$

idea: cancel ω_z w/ ω_z . Replace w/ ω_4 @ unity

Caveat if p/z misaligned $\rightarrow$ problem



	$p_1 \text{ (rad/s)}$	p_2	GBW	$p_3 \dots$	A_0'
Reduce A_0	$\frac{1}{R_1 C_1}$	$\frac{1}{R_2 C_2}$	$p_2 = \frac{1}{R_2 C_2}$		$A_0 \frac{\omega_1}{\omega_z}$
dom. pole	$\frac{1}{R_1 C_1'} \uparrow$	$\frac{1}{R_2 C_2}$			
p/z comp.	$\frac{1}{R_1 (C_1 + C_{p2})}$	$\frac{1}{R_{p2} C_{p2}} = \omega_z$	$\frac{1}{R_{p2} C_1} = \omega_4$		
Miller	$\uparrow$	$\downarrow$	$\frac{g_{m1}}{C_c}$		

Miller (sh-sh feedback)

if comp put across gain element, $z \uparrow$

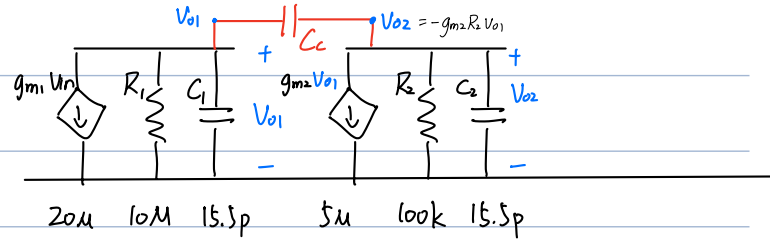
$$V_{cc} = V_{o1} - g_{m2} R_2 V_{o1} = (1 + g_{m2} R_2) V_{o1}$$

$$\omega_1 = \frac{1}{R_1 (C_1 + C_{eq})} = \frac{1}{R_1 A_2 C_c} \quad (\downarrow)$$

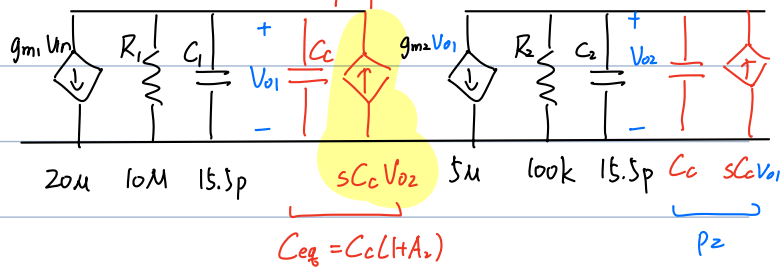
$$GBW = \frac{1}{R_1 A_2 C_c} (g_{m1} R_1 A_2) = \frac{g_{m1}}{C_c}$$

$$g_{m2eq} = g_{m2} - s C_c$$

$$\hookrightarrow \text{RHP } z @ \omega_{z2} = \frac{g_{m2}}{C_c}$$



$\downarrow$ y-param (sh-sh)



$$\frac{V_{o2}(s)}{V_{in}(s)} = \frac{g_{m1} R_1 g_{m2} R_2 (1 - \frac{s}{\omega_{z2}})}{1 + s [(C_1 + C_c (1 + g_{m2} R_2)) R_1] + s^2 [R_1 R_2 (C_2 + C_c) C_1 C_2 C_c]}$$

$$= \frac{N(s)}{1 + as + bs^2} = \frac{N(s)}{(1 + \frac{s}{\omega_1})(1 + \frac{s}{\omega_2})}$$

if $\omega_1 \ll \omega_2$ $\frac{1}{\omega_1} = R_1 (C_c (1 + g_{m2} R_2) + C_1) + R_2 (C_c + C_2)$

$$\omega_2 = \frac{R_1 (C_c (1 + g_{m2} R_2))}{R_1 R_2 (C_2 + C_c) C_1 + C_2 C_c} \approx \frac{g_{m2} C_c}{C_1 C_2 + C_1 C_c + C_2 C_c}$$

if $C_2 \gg C_c, C_2 \gg C_1$

$$\omega_1' \approx \omega_1 \frac{C_1}{g_{m2} R_2 C_c}$$

$$\omega_2' \approx g_{m2} R_2 \omega_2 \quad (C_2, C_c \gg C_1) \quad \uparrow$$

RHP z

$C_c = 0$

$$\frac{1}{R_1 C_1}$$

ω_1

$$\frac{1}{R_2 C_2}$$

ω_2

X

C_c large

$$\frac{1}{R_1 (C_1 + g_{m2} R_2 C_c)}$$

$$\frac{g_{m2}}{C_1 + C_2 + \frac{C_1 C_c}{C_c}}$$

$$\frac{g_{m2}}{C_c}$$

p_2 up !!! (pole splitting) $\rightarrow$ pm $\uparrow$

additional z LHP z $0^\circ \rightarrow +90^\circ$ help pm

RHP z $0^\circ \rightarrow -90^\circ$? hurt pm

leading 90°

lagging 90°

Need push z high, or make it LHP

If add large C directly, slow RC

but Miller, RC $\downarrow$

If add R_c , less overshoot / oscillation (improved PM)

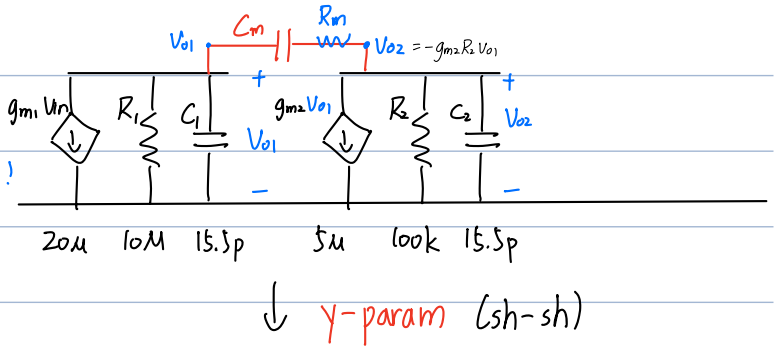
RHP zero compensation

Add $C_m - R_m$, avoid C_c feedfwd ($sC_c V_{o1}$)

$$g_{m2, eq} = g_{m2} - \frac{sC_m}{1+sR_{in}C_m} = g_{m2} \frac{1+sC_m(R_{in}-1/g_{m2}) \rightarrow 2!}{1+sR_{in}C_m}$$

If $R_M > \frac{1}{g_{m2}} \rightarrow$ no RHP z

(LHP helps)



V_{cm} is a new state var. $\rightarrow 3p$

if $g_{m1}R_1, g_{m2}R_2 \gg 1, C_m \gg C_1, C_2 \gg C_1$

$$W_{p1} \approx \frac{R_1 g_{m2} R_2 C_m}{g_{m2} C_m} \quad \left. \vphantom{\frac{R_1 g_{m2} R_2 C_m}{g_{m2} C_m}} \right\} \text{same}$$

$$W_{p2} \approx \frac{g_{m2} C_m}{C_1 C_2 + C_1 C_m + C_2 C_m}$$

$$W_{p3} \approx \frac{1}{R_m C_1} \rightarrow \text{new}$$

